

# Normally – OFF Silicon Carbide Junction Transistor

 $V_{DS}$  = 600 V  $R_{DS(ON)}$  = 25 mΩ  $I_{D (Tc = 25^{\circ}C)}$  = 100 A  $h_{FE (Tc = 25^{\circ}C)}$  = 105

#### **Features**

- 210°C maximum operating temperature
- · Gate Oxide Free SiC Switch
- Exceptional Safe Operating Area
- Excellent Gain Linearity
- Compatible with 5 V TTL Gate Drive
- Temperature Independent Switching Performance
- Low Output Capacitance
- Positive Temperature Coefficient of R<sub>DS,ON</sub>
- Suitable for Connecting an Anti-parallel Diode

### **Advantages**

- Compatible with Si MOSFET/IGBT Gate Drive ICs
- > 20 µs Short-Circuit Withstand Capability
- Lowest-in-class Conduction Losses
- High Circuit Efficiency
- Minimal Input Signal Distortion
- · High Amplifier Bandwidth

## **Package**





TO-258

## **Applications**

- Down Hole Oil Drilling
- Geothermal Instrumentation
- Solenoid Actuators
- General Purpose High-Temperature Switching
- Amplifiers
- Solar Inverters
- Switched-Mode Power Supply (SMPS)
- Power Factor Correction (PFC)

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## **Section I: Absolute Maximum Ratings**

Parameter	Symbol	Conditions	Value	Unit	Notes
Drain – Source Voltage	$V_{DS}$	V <sub>GS</sub> = 0 V	600	V	
Continuous Drain Current	I <sub>D</sub>	$T_J = 210^{\circ}C, T_C = 25^{\circ}C$	100	Α	
Continuous Gate Current	$I_{GM}$		3.5	Α	
Turn-Off Safe Operating Area	RBSOA	$T_J$ = 210°C, $I_G$ = 3.5 A, Clamped Inductive Load	$I_{D,max} = 50$ $\textcircled{0} V_{DS} \le V_{DSmax}$	Α	Fig. 18
Short Circuit Safe Operating Area	SCSOA	$T_J$ = 210°C, $I_G$ = 3.5 A, $V_{DS}$ = 400 V, Non Repetitive	>20	μs	
Reverse Gate – Source Voltage	$V_{SG}$	·	30	V	
Reverse Drain – Source Voltage	$V_{SD}$		25	V	
Power Dissipation	P <sub>tot</sub>	$T_J = 210^{\circ}C, T_C = 25^{\circ}C$	769	W	Fig. 16
Operating and Storage Temperature	T <sub>stg</sub>		-55 to 210	°C	



# **Section II: Static Electrical Characteristics**

Doromotor	Symbol	Canditions	Value		Linis	Natas	
Parameter		Conditions	Min.	Typical	Max.	- Unit	Notes
A: On State							
Drain – Source On Resistance	R <sub>DS(ON)</sub>	$\begin{array}{l} I_D = 50 \text{ A}, \ T_j = 25 \ ^{\circ}\text{C} \\ I_D = 50 \text{ A}, \ T_j = 125 \ ^{\circ}\text{C} \\ I_D = 50 \text{ A}, \ T_j = 175 \ ^{\circ}\text{C} \\ I_D = 50 \text{ A}, \ T_j = 210 \ ^{\circ}\text{C} \end{array}$		25 39 43 50		mΩ	Fig. 5
Gate – Source Saturation Voltage	$V_{GS,SAT}$	$I_D = 50 \text{ A}, I_D/I_G = 40, T_j = 25 ^{\circ}\text{C}$ $I_D = 50 \text{ A}, I_D/I_G = 30, T_j = 175 ^{\circ}\text{C}$		3.42 3.23		V	Fig. 7
DC Current Gain	h <sub>FE</sub>	$V_{DS} = 5 \text{ V}, I_D = 50 \text{ A}, T_J = 25 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_D = 50 \text{ A}, T_J = 125 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_D = 50 \text{ A}, T_J = 175 \text{ °C}$ $V_{DS} = 5 \text{ V}, I_D = 50 \text{ A}, T_J = 210 \text{ °C}$		105 77 71 70		_	Fig. 5
B: Off State							
Drain Leakage Current	I <sub>DSS</sub>	$\begin{array}{l} V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_J = 25 \text{ °C} \\ V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_J = 125 \text{ °C} \\ V_R = 600 \text{ V}, V_{GS} = 0 \text{ V}, T_J = 210 \text{ °C} \end{array}$		10 50 100		μA	Fig. 8
Gate Leakage Current	I <sub>sg</sub>	V <sub>SG</sub> = 20 V, T <sub>j</sub> = 25 °C		20		nA	
C: Thermal							
Thermal resistance, junction - case	$R_{thJC}$			0.26		°C/W	Fig. 19

# **Section III: Dynamic Electrical Characteristics**

Davamatan	Cumah al	Conditions		Value		1117	
Parameter	Symbol	Conditions	Min.	Typical	Max.	Unit	Notes
A: Capacitance and Gate Charg	е						
Input Capacitance	C <sub>iss</sub>	V <sub>GS</sub> = 0 V, V <sub>D</sub> = 100 V, f = 1 MHz		6450		pF	Fig. 9
Reverse Transfer/Output Capacitance	C <sub>rss</sub> /C <sub>oss</sub>	V <sub>D</sub> = 100 V, f = 1 MHz		420		pF	Fig. 9
Output Capacitance Stored Energy	Eoss	$V_{GS} = 0 \text{ V}, V_{D} = 400 \text{ V}, f = 1 \text{ MHz}$		17.4		μJ	Fig. 10
Effective Output Capacitance, time related	$C_{oss,tr}$	$I_D$ = constant, $V_{GS}$ = 0 V, $V_{DS}$ = 0400 V		390		pF	
Effective Output Capacitance, energy related	C <sub>oss,er</sub>	V <sub>GS</sub> = 0 V, V <sub>DS</sub> = 0400 V		284		pF	
Gate-Source Charge	Q <sub>GS</sub>	V <sub>GS</sub> = -53 V		55		nC	
Gate-Drain Charge	$Q_{GD}$	V <sub>GS</sub> = 0 V, V <sub>DS</sub> = 0400 V		156		nC	
Gate Charge - Total	$Q_{G}$			211		nC	
B: Switching <sup>1</sup>							
Internal Gate Resistance – zero bias	$R_{G(INT-ZERO)}$	$f = 1 \text{ MHz}, V_{AC} = 50 \text{ mV}, V_{DS} = 0 \text{ V}, V_{GS} = 0 \text{ V}, T_1 = 210 ^{\circ}\text{C}$		0.9		Ω	
Internal Gate Resistance – ON	$R_{G(INT-ON)}$	$V_{GS} > 2.5 \text{ V}, V_{DS} = 0 \text{ V}, T_j = 210 ^{\circ}\text{C}$		0.09		Ω	
Turn On Delay Time	$t_{d(on)}$	$T_i = 25 ^{\circ}\text{C},  V_{DS} = 400 ^{\circ}\text{V},$		25		ns	
Fall Time, V <sub>DS</sub>	t <sub>f</sub>	_I <sub>D</sub> = 50 A, Resistive Load		44		ns	Fig. 11,13
Turn Off Delay Time	$t_{d(off)}$	Refer to Section V for additional		40		ns	
Rise Time, V <sub>DS</sub>	t <sub>r</sub>	driving information.		33		ns	Fig. 12,14
Turn On Delay Time	$t_{d(on)}$			19		ns	
Fall Time, V <sub>DS</sub>	t <sub>f</sub>	T <sub>i</sub> = 210 °C, V <sub>DS</sub> = 400 V,		43		ns	Fig. 11
Turn Off Delay Time	$t_{d(off)}$	I <sub>D</sub> = 50 A, Resistive Load		89		ns	
Rise Time, V <sub>DS</sub>	t <sub>r</sub>	_		27		ns	Fig. 12
Turn-On Energy Per Pulse	Eon	_T <sub>i</sub> = 25 °C, V <sub>DS</sub> = 400 V,		690		μJ	Fig. 11,13
Turn-Off Energy Per Pulse	E <sub>off</sub>	I <sub>D</sub> = 50 A, Inductive Load		359		μJ	Fig. 12,14
Total Switching Energy	E <sub>tot</sub>	Refer to Section V.		1049		μJ	
Turn-On Energy Per Pulse	Eon	T 04000 14 40014		758		μJ	Fig. 11
Turn-Off Energy Per Pulse	E <sub>off</sub>	─T <sub>j</sub> = 210 °C, V <sub>DS</sub> = 400 V, —I <sub>D</sub> = 50 A, Inductive Load		337		μJ	Fig. 12
Total Switching Energy	E <sub>tot</sub>	—ip - 50 A, illuuctive Loau		1095		μJ	-

 $<sup>^{\</sup>rm 1}-$  All times are relative to the Drain-Source Voltage  $V_{\rm DS}$ 



# **Section IV: Figures**

## A: Static Characteristics

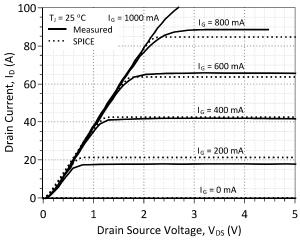


Figure 1: Typical Output Characteristics at 25 °C

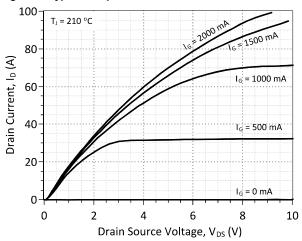


Figure 3: Typical Output Characteristics at 210 °C

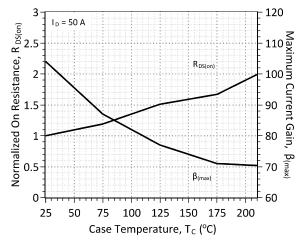


Figure 5: Normalized On-Resistance and Current Gain vs. Temperature

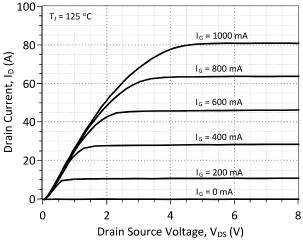


Figure 2: Typical Output Characteristics at 125 °C

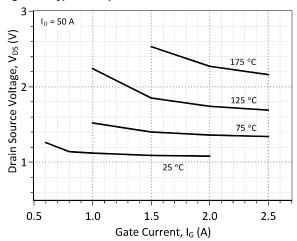


Figure 4: Drain-Source Voltage vs. Gate Current

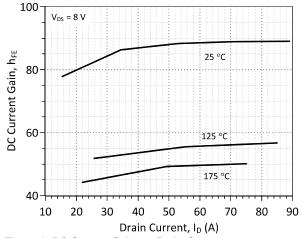


Figure 6: DC Current Gain vs. Drain Current

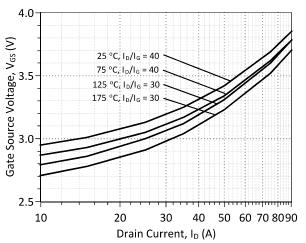
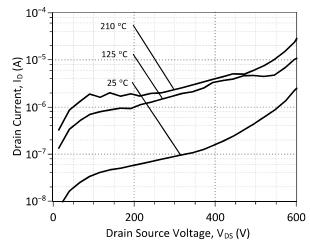


Figure 7: Typical Gate - Source Saturation Voltage



**Figure 8: Typical Blocking Characteristics** 

## **B: Dynamic Characteristics**

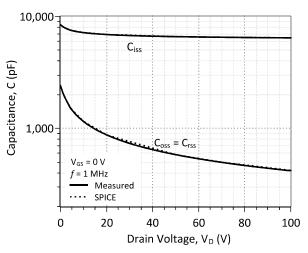


Figure 9: Input, Output, and Reverse Transfer Capacitance

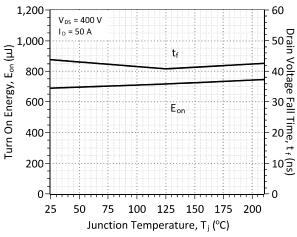


Figure 11: Typical Turn On Energy Losses and Switching Times vs. Temperature

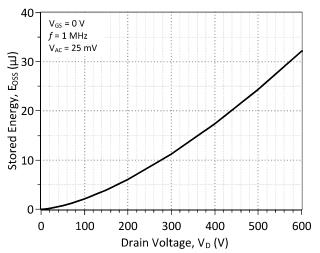


Figure 10: Output Capacitance Stored Energy

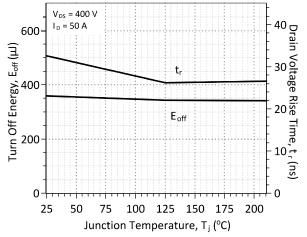


Figure 12: Typical Turn Off Energy Losses and Switching Times vs. Temperature





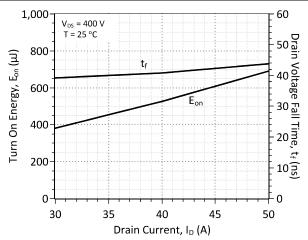
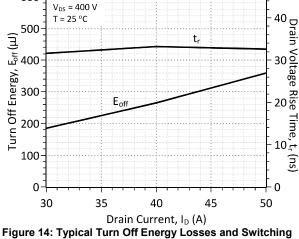


Figure 13: Typical Turn On Energy Losses and Switching Times vs. Drain Current



600

Times vs. Drain Current

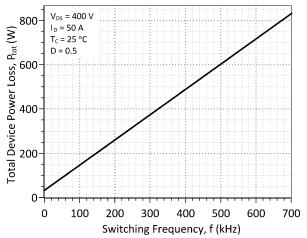


Figure 15: Typical Hard Switched Device Power Loss vs. Switching Frequency

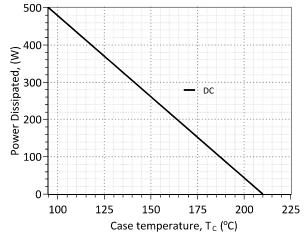


Figure 16: Power Derating Curve

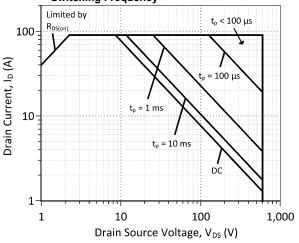


Figure 17: : Forward Bias Safe Operating Area at T<sub>c</sub>= 25 °C

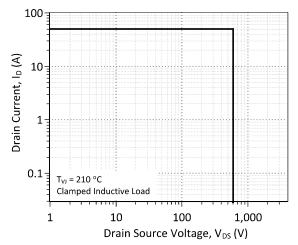


Figure 18: Turn-Off Safe Operating Area

<sup>&</sup>lt;sup>2</sup> – Representative values based on device conduction and switching loss. Actual losses will depend on gate drive conditions, device load, and circuit topology.

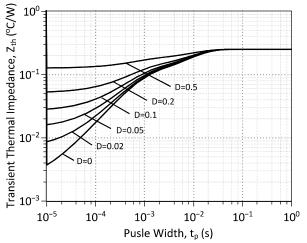


Figure 19: Transient Thermal Impedance

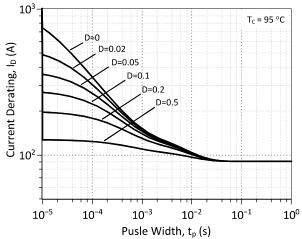


Figure 20: Drain Current Derating vs. Pulse Width



## Section V: Driving the GA50JT06-247

The GA50JT06-247 is a current controlled SiC transistor which requires a positive gate current for turn-on and to remain in on-state. It may be driven by different drive topologies depending on the intended application.

Table 1: Estimated Power Consumption and switching frequencies for various Gate Drive topologies.

Drive Topology	Gate Drive Power Consumption	Switching Frequency
Simple TTL	High	Low
Constant Current	Medium	Medium
High Speed – Boost Capacitor	Medium	High
High Speed – Boost Inductor	Low	High
Proportional	Lowest	Medium
Pulsed Power	Medium	N/A

#### A: Simple TTL Drive

The GA50JT06-247 may be driven by 5 V TTL logic by using a simple current amplification stage. The current amplifier output current must meet or exceed the steady state gate current,  $I_{G,steady}$ , required to operate the GA50JT06-247. An external gate resistor  $R_G$ , shown in the Figure 21 topology, sets  $I_{G,steady}$  to the required level which is dependent on the SJT drain current  $I_D$  and DC current gain  $h_{FE}$ ,  $R_G$  may be calculated from the equation below. The values of  $h_{FE}$  and  $V_{GS,sat}$  may be read from Figure 6 and Figure 7, respectively.  $V_{EC,sat}$  can be taken from the PNP datasheet, a partial list of high-temperature PNP and NPN transistors options is given below. High-temperature MOSFETs may also be used in the topology.

$$R_{G,max} = \frac{\left(5.0 \ V - V_{EC,sat}(PNP) - V_{GS,sat}(SJT)\right) * h_{FE}(T, I_D)}{I_D * 1.5}$$

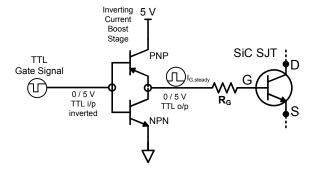


Figure 21: Simple TTL Gate Drive Topology

Table 2: Partial List of High-Temperature BJTs for TTL Gate Driving

BJT Part Number	Туре	T <sub>j,max</sub> (°C)
PHPT60603PY	PNP	175
PHPT60603NY	NPN	175
2N2222	NPN	200
2N6730	PNP	200
2N2905	PNP	200
2N5883	PNP	200
2N5885	NPN	200



#### **B: High Speed Driving**

For ultra high speed GA50JT06-247 switching ( $t_r$ ,  $t_f < 20$  ns) while maintaining low gate drive losses the supplied gate current should include a positive current peak during turn-on, a negative voltage peak during turn-off, and continuous gate current I<sub>G</sub> to remain on.

An SJT is rapidly switched from its blocking state to on-state, when the necessary gate charge for turn-on,  $Q_G$ , is supplied by a burst of high gate current until the gate-source capacitance,  $C_{GS}$ , and gate-drain capacitance,  $C_{GD}$ , are fully charged. Ideally, the burst should terminate when the drain voltage has fallen to its on-state value in order to avoid unnecessary drive losses. A negative voltage peak is recommended for the turn-off transition in order to ensure that the gate current is not being supplied under high dV/dt due to the Miller effect. While satisfactory turn off can be achieved with  $V_{GS} = 0$  V, a negative  $V_{GS}$  value may be used in order to speed up the turn-off transition.

#### B:1: High Speed, Low Loss Drive with Boost Capacitor

The GA50JT06-247 may be driven using a High Speed, Low Loss Drive with Boost Capacitor topology in which multiple voltage levels, a gate resistor, and a gate capacitor are used to provide current peaks at turn-on and turn-off for fast switching and a continuous gate current while in on-state. As shown in Figure 22, in this topology two gate driver ICs are utilized. An external gate resistor R<sub>G</sub> is driven by a low voltage driver to supply the continuous gate current throughout on-state.and a gate capacitor C<sub>G</sub> is driven at a higher voltage level to supply a high current peak at turn-on and turn-off. A 3 kV isolated evaluation gate drive board (GA03IDDJT30-FR4) from GeneSiC Semiconductor utilizing this topology is commercially available for high and low-side driving, its datasheet provides additional details about this drive topology.

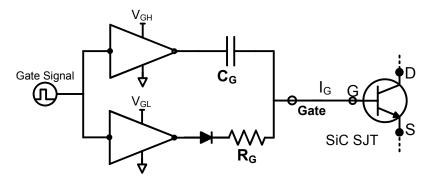


Figure 22: High Speed, Low Loss Drive with Boost Capacitor Topology

#### B:2: High Speed, Low Loss Drive with Boost Inductor

A High Speed, Low-Loss Driver with Boost Inductor is also capable of driving the GA50JT06-247 at high-speed. It utilizes a gate drive inductor instead of a capacitor to provide the high-current gate current pulses  $I_{G,on}$  and  $I_{G,off}$ . During operation, inductor L is charged to a specified  $I_{G,on}$  current value then made to discharge  $I_L$  into the SJT gate pin using logic control of  $S_1$ ,  $S_2$ ,  $S_3$ , and  $S_4$ , as shown in Figure 23. After turn on, while the device remains on the necessary steady state gate current  $I_{G,steady}$  is supplied from source  $V_{CC}$  through  $R_G$ . Please refer to the article "A current-source concept for fast and efficient driving of silicon carbide transistors" by Dr. Jacek Rąbkowski for additional information on this driving topology.<sup>3</sup>

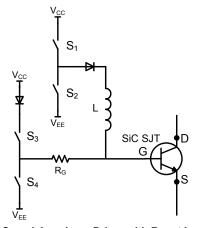


Figure 23: High Speed, Low-Loss Driver with Boost Inductor Topology

<sup>&</sup>lt;sup>3</sup> – Archives of Electrical Engineering. Volume 62, Issue 2, Pages 333–343, ISSN (Print) 0004-0746, DOI: 10.2478/aee-2013-0026, June 2013



#### C: Proportional Gate Current Driving

A proportional gate drive topology may be beneficial for applications in which the GA50JT06-247 will operate over a wide range of drain current conditions to lower the gate drive power consumption. A proportional gate driver relies on instantaneous drain current  $I_D$  feedback to vary the steady state gate current  $I_{G,steady}$  supplied to the GA50JT06-247.

#### C:1: Voltage Controlled Proportional Driver

A voltage controlled proportional driver relies on a gate drive integrated circuit to detect the GA50JT06-247 drain-source voltage  $V_{DS}$  during on-state to sense  $I_D$ . The integrated circuit will then increase or decrease  $I_G$  in response to  $I_D$ . This allows  $I_G$  and gate drive power consumption to reduce while  $I_D$  is low or for  $I_G$  to increase when  $I_D$  increases. A high voltage diode connected between the drain and sense protects the integrated circuit from high-voltage when blocking. A simplified version of this topology is shown in Figure 24. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

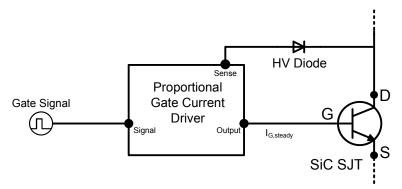


Figure 24: Simplified Voltage Controlled Proportional Driver

## C:2: Current Controlled Proportional Driver

The current controlled proportional driver relies on a low-loss transformer in the drain or source path to provide feedback of the GA50JT06-247 drain current during on-state to supply  $I_{G,steady}$  into the gate.  $I_{G,steady}$  will increase or decrease in response to  $I_D$  at a fixed forced current gain which is set be the turns ratio of the transformer,  $h_{force} = I_D / I_G = N_2 / N_1$ . GA50JT06-247 is initially tuned-on using a gate current pulse supplied into an RC drive circuit to allow  $I_D$  current to begin flowing. This topology allows  $I_{G,steady}$  and the gate drive power consumption to reduce while  $I_D$  is relatively low or for  $I_{G,steady}$  to increase when  $I_D$  increases. A simplified version of this topology is shown in Figure 25. Additional information will be available in the future at http://www.genesicsemi.com/references/product-notes/.

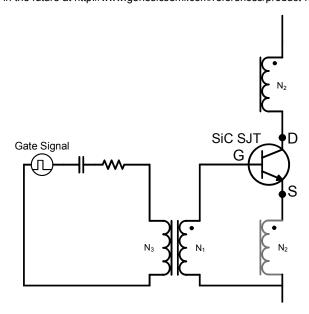
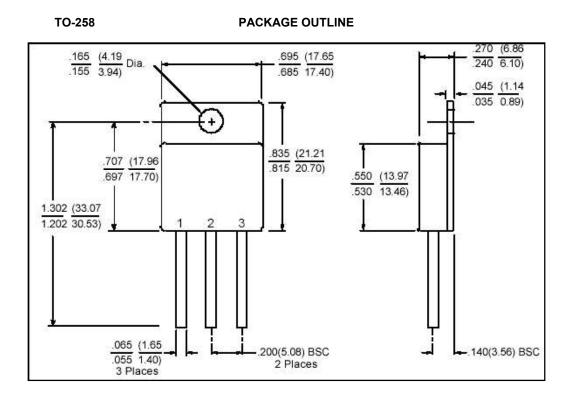


Figure 25: Simplified Current Controlled Proportional Driver



## **Section VI: Package Dimensions**



#### NOTE

- 1. CONTROLLED DIMENSION IS INCH. DIMENSION IN BRACKET IS MILLIMETER.
- 2. DIMENSIONS DO NOT INCLUDE END FLASH, MOLD FLASH, MATERIAL PROTRUSIONS

Revision History					
Date	Revision	Comments	Supersedes		
2014/12/12	5	Updated Electrical Characteristics			
2014/08/23	4	Updated Electrical Characteristics			
2014/04/10	3	Updated Electrical Characteristics			
2014/02/05	2	Updated Electrical Characteristics			
2013/12/19	1	Updated Gate Drive Section			
2013/12/05	0	Initial release			

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## **Section VII: SPICE Model Parameters**

This is a secure document. Please copy this code from the SPICE model PDF file on our website (http://www.genesicsemi.com/images/hit\_sic/sjt/GA50JT06-258\_SPICE.pdf) into LTSPICE (version 4) software for simulation of the GA50JT06-258.

```
MODEL OF GeneSiC Semiconductor Inc.
     $Revision: 1.3
     $Date: 12-DEC-2014
                                   $
     GeneSiC Semiconductor Inc.
     43670 Trade Center Place Ste. 155
     Dulles, VA 20166
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* These models are provided "AS IS, WHERE IS, AND WITH NO WARRANTY
* OF ANY KIND EITHER EXPRESSED OR IMPLIED, INCLUDING BUT NOT LIMITED
* TO ANY IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A
* PARTICULAR PURPOSE."
* Models accurate up to 2 times rated drain current.
.model GA50JT06 NPN
+ IS
           5.00E-47
+ ISE
          1.26E-26
+ EG
           3.23
          106
+ BF
+ BR
          0.55
          9000
+ IKF
           1
+ NF
          2
+ NE
          0.9
+ RB
+ IRB
          0.002
+ RBM
          0.09
          0.01
+ RE
          0.013
+ RC
+ CJC
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           2.8346223
+ VJC
+ MJC
          0.4846
+ CJE
          6.026E-09
          3.17915435
+ VJE
+ MJE
          0.52951635
+ XTI
+ XTB
          -1.2
+ TRC1
                 7.00E-3
                 600
+ VCEO
+ ICRATING 100
+ MFG GeneSiC Semiconductor
```

\* End of GA50JT06 SPICE Model